

Fractional Order PI Control in Hybrid Renewable Power Generation System to Three Phase Grid Connection

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Abstract: The main objective of this paper is to allow renewable energy sources (RES) to actively participate within hybrid microgrid by proposing a new control system based on fractional order proportional integral (FOPI) controller. Fractional order proportional integral controller is a classical proportional integral (PI) in which the integral part is a fraction instead of integer numbers. The paper introduces a hybrid photovoltaic (PV), wind turbine and battery storage system connected to a three-phase grid. Three types of controller are considered and compared for a hybrid renewable energy system (HRES), namely, FOPI, PI, and the fractional order integral control (FIC). For the PV resource, maximum power point tracking (MPPT) controller was designed using the incremental conductance plus integral regulator technique. A DC/DC boost converter was utilized to connect the renewable energy resources to a point of common coupling. MATLAB/Simulink is adopted to perform the simulation results of the developed HRES. The results show that the FOPI controller outperforms other controllers under several operating conditions. The paper also includes experimental results from a prototype real scale.

Keywords: photovoltaic, fractional order proportional integral, hybrid renewable energy system, maximum power point tracking, voltage source control, three-phase grid.

1. Introduction

Because of environmental problems and energy crises, clean and renewable energy generation resources are important priorities by many governments around the world and among the most widespread renewable energy are solar and wind energy [1]. Renewable energy sources (RES) have been filtered to solve problems related to the increasing demand for electrical loads and the environmental side effects of traditional fossil fuel sources that cause numerous damages to humans and society, which are mainly used in conventional power plants [2]. The increasing trend of renewable energy is in response to the harmful environmental impacts of increased carbon emissions for conventional power plants and increased electrical loads worldwide [3], [4].

The most common components in hybrid renewable energy systems (HRESs) are the PV, wind turbine and battery storage. Normally, these three types of resources are connected together via the DC-DC converters. A DC-DC converter transforms the DC energy to an appropriate DC level at the point of common coupling, which is a compulsory step for grid connected AC three-phase inverters. The HRES combines several renewable resources, which is deemed to be a solution to avoid the problems of dependence on a single resource [5] to meet load demands or transfer such renewable energy to a three-phase grid. All these resources are connected in parallel to supply the energy demands [6]. Normally, maximum power point tracking techniques (MPPT) are employed to significantly enhance the efficiency of the PV modules and wind-turbine systems [7]. However, there has been an increase in the number of resources which are DC in nature, or whose ac frequency is neither constant nor match the utility grid frequency. In some of these cases, three-phase inverters have been used. Besides, a control of both active and reactive powers is provided in order to deliver the electrical energy with the same utility grid frequency [8].

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With the above challenges, the authors of this work are motivated to develop modern controllers thereby regulating the energy transferred by voltage source inverter (VSI). In its many applications of fractional differentiation and integration, science and engineering branches use many fractional derivatives. The most widely used in research are Caputo and Riemann-Liouville derivatives, which have been widely used in modeling complex dynamics that appear in physics, biology, engineering, and many other fields [9].

The fractional order proportional integral (FOPI or PI λ) controller has been a topic of interest of researchers from academia and industry in the last years. They are more flexible and provide improved performance than conventional PI controllers [10] rather than the basic characteristics of the PI adapted on the power systems [11,12]. In the literature, the fractional-order PI (FOPI) and integral controllers were being investigated. In [13], F. K. Abo-Elyousr et al (2019) has developed an FOPI controller, which renders multiple poles and zeros in the designing procedures, to solve load frequency issues regarding massless inertia resources. The authors of [13] demonstrated that FOPI controller had outperformed the other controllers. In another research [14], A. Kumar (2020) et al utilized a fractional derivative with particle swarm optimization to design a finite impulse response. V. A. Kumar et al (2020) have investigated fractional order controller for cascaded islanded H-bridge [15]. In another article [16], fractional order PID was introduced to enhance the stability of HVDC system. A. Narendran et al (2020) conducted a research to design an inverter fed induction motor whereby the fractional order controller demonstrated a competitiveness amongst the other controllers [17]. A. Rawat et al (2020) conducted a research to employ FOPID for PVs maximum power point tracking [18]. In another research [19], a fractional order PI was utilized to control the pitch angle of wind turbines. B. Puschalski et al (2020) conducted a research via fraction order PID together with fuzzy controller for a pressurized water type nuclear reactor with a conclusion that fractional order controllers have strong efficacy [20]. The investigation of the above manuscripts reveals that fractional order controllers are limited in the literature regarding HRES based grid connected inverters, which gives the authors another motivation to develop fractional order-based controllers.

At the level of servo systems, Z. Qi et al (2019) conducted a research to design a FOPID and concluded that FOPID has better performance [21]. In another article [22], F. Meng et al (2020) introduced fractional order PID to control a motor torque and demonstrated that fractional order controllers had better performance. A survey for fractional order controllers was conducted in [23]. A design algorithm for fractional controllers combined with fuzzy was conducted by J. Z. Shi with a conclusion that fractional controllers had proven better performance (2020) [24]. Fractional order was utilized by Zaheeruddin et al (2020) at the level of hybrid microgrid to design a load frequency controller [25]. In another article [26], R. P. Kumar et al (2020) introduced a FACT based on fractional order PID with PQ coordination theory for a grid connected RES and showed that FOPID had surpassed the other developed controllers regarding the power quality issues. In [27], fractional order controllers were utilized for an air heating system. In [28], fractional order PID controller based fuzzy was employed to fix to mitigate the vibration isolation for a meteorological elastomer. A. K. Patra et al. (2020) introduced a fractional order PID for a suspension vehicle [29]. Fractional order controllers were adopted by M. S. Bouakkaz to help enhance the PV generation [30].

Despite the rich literature survey of fractional order in several fields, yet, the application of fractional order controllers for grid connected HRES is rare. For such challenging tasks, the paper contributes in the literature as (i) to develop a fractional order PI (FOPI) controller, (ii) the effectiveness developed FOPI is verified through a fair comparison with the classical and integral order PI, and fractional order integral controllers, (ii) considering the changeable conditions of PV (irradiance and temperature) and wind turbine (wind speed) of the system.

The remaining of this paper is organized as follows, system description and proposed controllers are introduced in sections 2 and 3. The simulation results and discussions are given in section 4. The experimental results and conclusions are given in sections 5 and 6, respectively.

2. System Description

The HERS generation system proposed in this paper as shown in Figure 1. The proposed system consists of a wind generation, PV system, energy storage, boost converter, DC link, three-phase inverter, controller, high transformer, and the utility grid and AC loads. Normally, a VSI inverter is employed to change form the DC side to the grid the utility grid side.

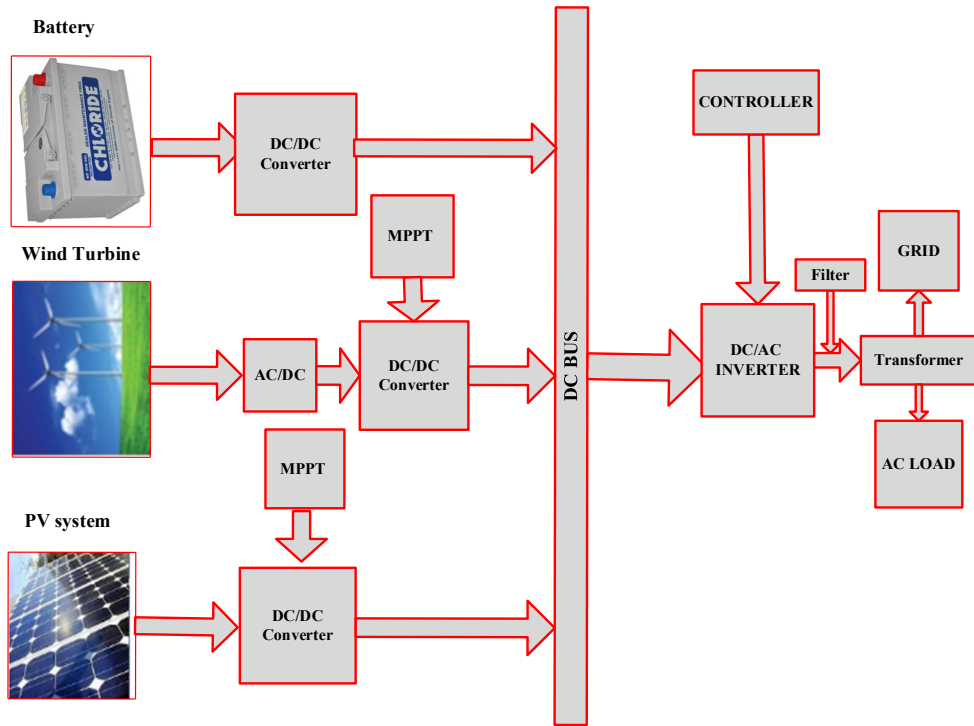


Figure 1. Hybrid renewable energy generation system configuration.

A. PV System

Electrical Energy is used to customers, Converters are used to convert the PV energy which come from sun light into continuous electrical energy [3]. Modeling and simulation have a very important role for solar cells in designing true simulation of PV modules [30]-[31].

The equation of output voltage from PV describes the voltage root mean square (rms), which is being provided by the PV module voltage (V_{dc}).

$$V_{rms} = 0.612 M_a V_{dc} \quad (1)$$

Mathematical modeling of solar cells is an important step in the analysis and design of photovoltaic control systems. The photoelectric mathematical model can be obtained from the following equations: The equation that describes the “current-voltage” of the PV cell is given as:

$$I_c = I_{ph} - I_o \left\{ e^{\left[\frac{q}{AkT} (V_c + I_c R_s) \right]} - 1 \right\} \quad (2)$$

$$V_c = \frac{AkT}{q} \ln \left(\frac{I_{ph} + I_o - I_c}{I_o} \right) - I_c R_s \quad (3)$$

$$I = I_{ph} - I_o \left\{ e^{\left[\frac{q}{n_s A K T} (V + n_s I R_s) \right]} - 1 \right\} \quad (4)$$

$$V = \frac{n_s A K T}{q} \ln \left(\frac{I_{ph} + I_o - I}{I_o} \right) - n_s I R_s \quad (5)$$

Where;

$$I_{ph} = \frac{G}{1000} [I_{sc} + k_i (T - T_r)] \quad (5a)$$

$$I_o = I_{or} \left(\frac{T}{T_r} \right)^3 e^{\left[\frac{q E_g}{A K} \left(\frac{1}{T_r} - \frac{1}{T} \right) \right]} \quad (5b)$$

The module output power could be detected simply as in (6).

$$P = VI \quad (6)$$

The PV cell equivalent circuit is shown in Figure. 2. From the PV model, the I-V curve and P-V curve in Figure.3. The incremental conductance plus integral regulator technique is utilized to find the global MPPT of the PV module.

Where, P is output PV power, I is output PV current, V_c is cell PV voltage (V), I_o is diode current (A), I_{or} is diode saturation current (A), V is diode voltage (V), K is constant of boltzmann = 1.3806×10^{-23} J.K-1, n_1 is diode factor of ideality up to 1.0, q is charge of electron = 1.6022×10^{-19} , G is solar radiation and T is cell temperature (Co).

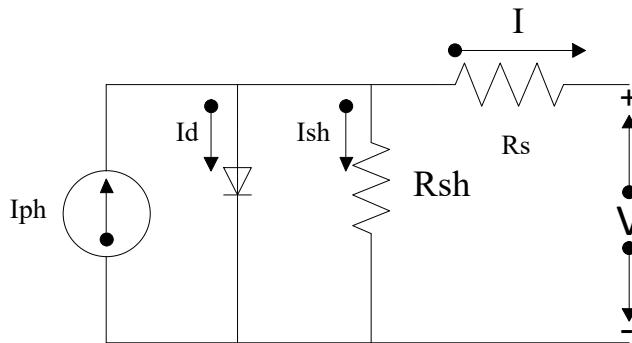


Figure 2. PV cell equivalent circuit.

B. Wind Turbine

The generator can optionally be a synchronous or asynchronous generator. Variable speed wind turbines are used with a full-range power transformer with a permanent magnetic synchronous generator (PMSG). PMSG is one of the most used types in wind energy because it is characterized by more gravity in wind energy applications because of its nature in using permanent magnets to obtain magnetic field instead of winding, which causes copper losses [32]. This configuration is compatible with the changing wind turbines with full speed generator connected to the network through a full-scale frequency converter [33]. Wind speed is the important factor for determining the power of wind turbines. In this research was used PMSG [34]. For system design and control, the need for dynamic modeling of a wind energy conversion system is required in this paper. Wind energy is obtained from the kinetic energy (E) given by equation (7). In order for wind energy to be obtained by distinguishing kinetic energy in the wind as shown in equation (8). In equation (9), the wind mass flow rate for obtaining electrical energy is discribed. By replacing equations (9) in (8), wind energy can be expressed as in (10).

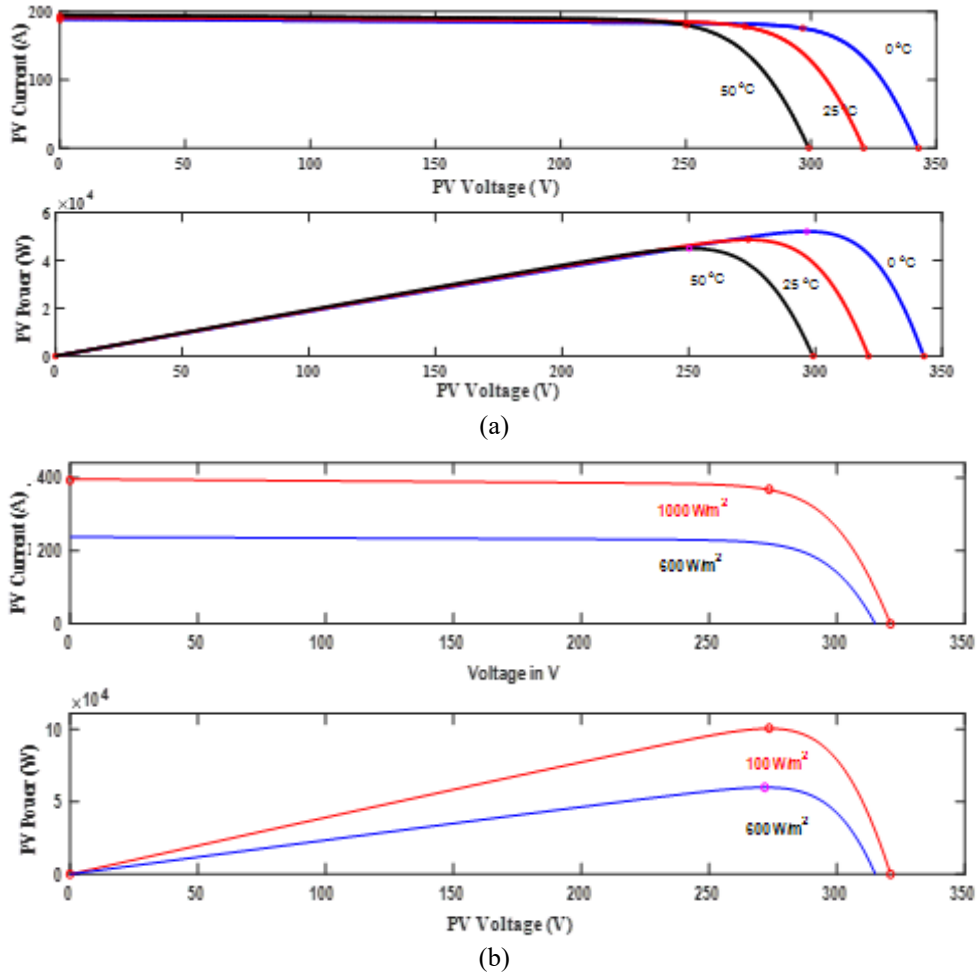


Figure 3. PV output characteristics I-V and P-V curves at (a) different temperature, (b) different irradiation.

$$E = \frac{1}{2} \dot{m} * V^2 \quad (7)$$

$$P = \frac{dE}{dt} = \frac{1}{2} \dot{m} * V^2 \quad (8)$$

$$\dot{m} = \rho A v \quad (9)$$

$$p = 0.5 \rho A v^3 \quad (10)$$

The calculation of the actual mechanical capacity to generate the wind energy harvested from wind turbines can be obtained as in equation (11). Besides, the mechanical torque is given as in (12).

$$P_m = \frac{1}{2} \rho \pi R^2 C_p V^3 \quad (11)$$

$$T_m = \frac{P_m}{\omega_m} \quad (12)$$

Most PMSG properties are similar to the synchronous generator used in wind energy generation. The mathematical model equation voltages of the machine are calculated in (13).

$$V_{dq0} = K_{cp} V_{abc} \quad (13)$$

The tip speed ratio of a wind turbine can be calculated in (14). Figure 4 shows a typical C_p versus λ_r curve for PMSG of wind turbine.

$$\lambda_r = \frac{w_m R}{V_w} \quad (14)$$

Where; V_d, V_q represents the stator voltages in the d-q axis respectively, ω_s is the angular velocity of the synchronously rotating reference frame, C_p is the power coefficient, P_m is mechanical power, $\dot{m}$ is the air mass, v is the wind speed, ρ is the density of air, A is the rotor blade area and R the wind turbine rotor radius. The C_p is related to and tip speed ratio (λ_r), T_m is mechanical torque (N.m), VW is the wind speed (m/s), β is pitch angle of blades (deg), ω_r is the wind turbine rotational speed (rad/s).

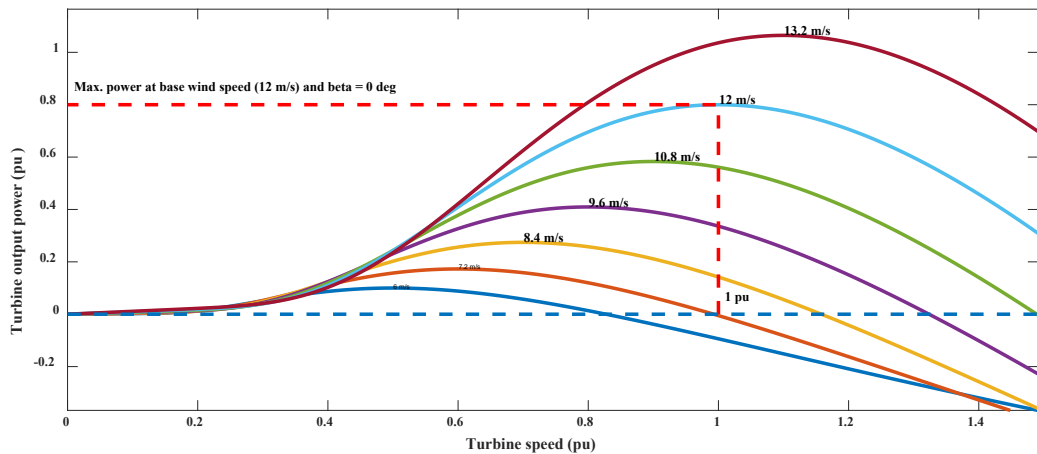


Figure 4. A typical C_p versus λ_r curve.

C. Battery Storage

Battery storage is important part of the system HRES. Electrical energy generated from renewable sources depends entirely on weather conditions such as wind speed and variable sunlight throughout the day, which enhances the use of energy storage necessary with it. [35]. Power Storage is the key concerns worldwide when developing the HRES. If conditions are bad the power from the system to grid is low, here battery play important role to give the energy [36]. Herein, Ni-MH connected to a DC-DC boost converter is utilized. Because intermittent nature of renewables, a battery storage is utilized to meet the power demand to the grid and the local loads.

Electric energy generated from renewable sources depends on changing weather conditions such as wind speed and sunlight. Power Storage is the key concerns worldwide when developing the HRES [28]. PV panels work in sun light at all days and wind turbine work in all the best of time the conditions allow to it. Because intermittent nature of renewables, A battery storage is utilized to get the power demand to the grid and the loads connect with the system. The type of the used battery is Ni-MH battery connected to a DC-DC boost converter.

D. DC-DC Converter

In this system authors used the type of DC-DC Converter is Boost converter to set up voltage level. The maximum power from the HRES voltage level is raised to a suitable level, then transferred to the three-phase inverter and to the grid or AC load through DC-DC converter [37]. The output voltage can be expressed as the next equation. the topology of boost converter as shown in Figure 5.

$$V_o = \frac{V_s}{1-D} \quad (15)$$

Where, V_s is converter input voltage (V), V_o is boost converter output voltage (V) and D is duty cycle of the boost converter.

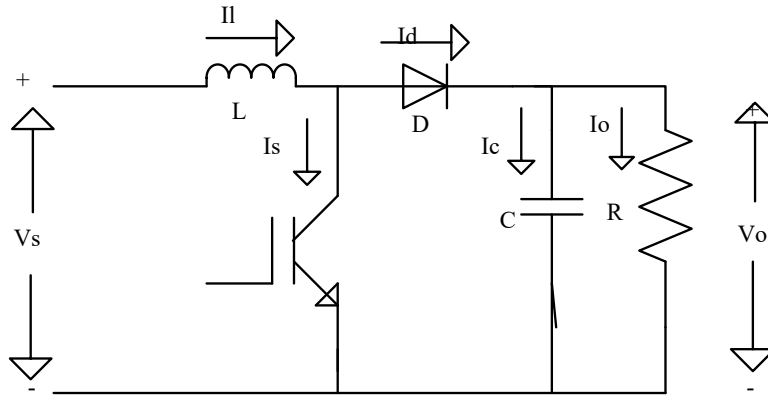


Figure 5. DC-DC Boost Converter.

3. Proposed controllers

A new approach is used to control the three-way switch to improve the performance of the networked HRES system. As shown in Figure. 6, it uses four steps at the same time of control: the first control loop, which is a process that regulates the AC voltage which controls the size of the AC voltage from VSI, and the second is the constant voltage of the capacitor DC which is the DC voltage regulator which is the preservation On the output voltage of VSI. The third and fourth controls for regulating the volume and phase of the VSI output current at appropriate values under different operating conditions in the wind and solar radiation to maintain the inverter output power factor in the unit operation. [38].

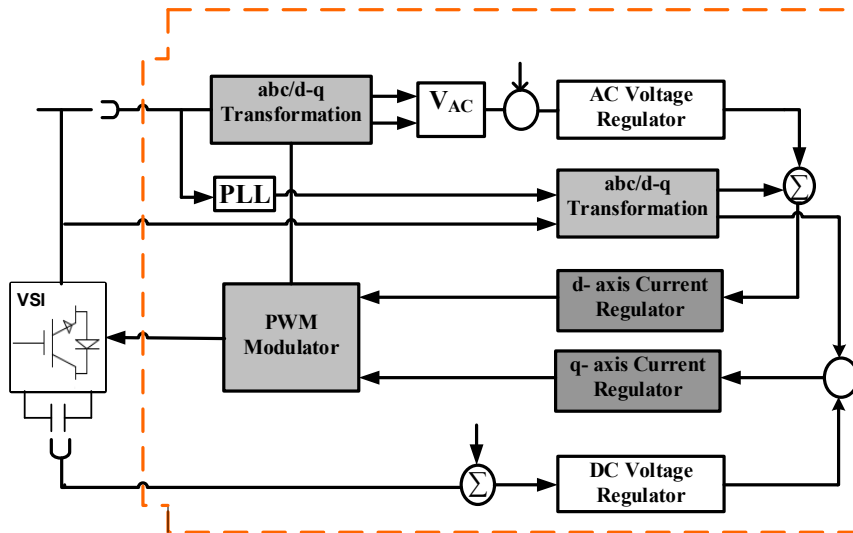


Figure 6. Block of controller in inverter of the system.

A. Classic Proportional Integral Controller (PI)

PI control is still the most popular feedback design in industry. The PI controller has two constant parameters K_p for proportional and K_i for integral term [39]. The operating logic of the PI controller considers the optimum technique to be used in HRES. The summation of two terms adjusts the process through a control element as in (16).

$$G(s) = K_p + \frac{K_i}{s} \quad (16)$$

Where, K_p is a proportional control, K_i is an integral control and $G(s)$, is the functions of the variable s .

The goal of the integrated terminology of the PI controller is to have a stable state error of zero for step input in long-term operating conditions [40], [41]. Figure. 7 demonstrates a schematic diagram of the internal structure of the PI block.

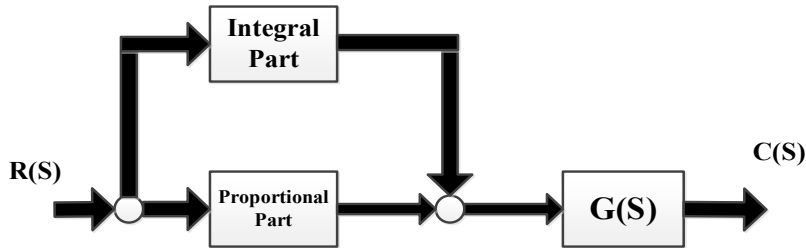


Figure 7. Internal structure of PI block

B. Integral Control (IC)

The IC parameters depend on non-integer-order fundamental operator, aD_t^α , where a and t are the bounds of operation and $\alpha \in \mathbb{R}$ [42]. The continuous integral operation is defined as (17).

$$aD_t^\alpha = \begin{cases} \frac{d^\alpha}{dt^\alpha} & : \alpha > 0, \\ 1 & : \alpha = 0, \\ \int_a^t (dT) & : \alpha < 0. \end{cases} \quad (17)$$

Where, aD_t^α is a generalization of integration and differentiation to non-integer-order fundamental operator, the components a and t are bounds of operation. The important function used in fractional integral is called gamma function, which is defined as:

$$\Gamma(n) = \int_0^\infty t^{n-1} e^{-t} dt. \quad (18)$$

This function is a generalization of the factorial in the following from:

$$\Gamma(n) = (n-1)! \quad (19)$$

Where, Γ is the gamma function is directly implemented in MATLAB with syntax

C. Fractional Order PI Control (FOPI)

From the control methods, fractional order PI^λ controller was introduced as a generalization of the classical PI controller [42]. Generally, FOPI is a classic PI whereby the integral order of the Laplace's operator is a fraction rather than integer.

Several partial differentiation definitions have been suggested thereby the most common among them is the definition of Riemann-Liouville (R-L), the definition of Grunwald-Letnikov (G-L) [43]. The following equations show the basics of fractional calculus.

1- The Riemann-Liouville (R-L) definition is given as in (20).

$$aD_t^\alpha f(t) = \frac{1}{\Gamma(n-\alpha)} \frac{d^n}{dt^n} \int_a^t \frac{f(\tau)}{(t-\tau)^{(\alpha-n+1)}} d\tau, \quad (20)$$

Where, $f(t)$ is a function assumed to be sufficiently smooth and locally integrable and $\Gamma(\cdot)$ is the gamma function for $(n-1 < \alpha < n)$.

2- The Grunwald-Letnikov (G-L) definition

$$aD_t^\alpha f(t) = \lim_{h \rightarrow 0} h^{-\alpha} \sum_{j=0}^{\left[\frac{t-a}{h}\right]} (-1)^j \alpha C_j f(t - jh), \quad (21)$$

where, αC_j is a binomial coefficient.

3- The Caputo definition is given as

$$aD_t^\alpha f(t) = \frac{1}{\Gamma(n-\alpha)} \int_a^t \frac{f^n(\tau)}{(t-\tau)^{(\alpha-n+1)}} d\tau. \quad (22)$$

Among the aforementioned three definitions, the most commonly used is the Riemann-Liouville (R-L) definition. The R-L definition and Caputo definition are almost the same except for some initial value settings [44, 45]. The Laplace transform under zero initial conditions can be expressed as in (23) [13], [42].

$$L\{aD_t^{\pm\alpha} f(t)\} = s^{\pm\alpha} \cong s^\lambda F(s), \quad 0 < \lambda < 1 \quad (23)$$

The difference between PI control and FOPI control is λ , which is a fraction rather than integer. Provided that λ is positive for differentiation and negative for integration, FOPI control is calculated as in (24).

$$G(s) = \frac{u(s)}{e(s)} = K_p + \frac{K_i}{s^\lambda} \quad (24)$$

where, $G(s)$, $u(s)$ and $e(s)$ are rational functions of the variable s , K_p is a proportional control and K_i is an integral control.

4. Simulation Results and Discussions

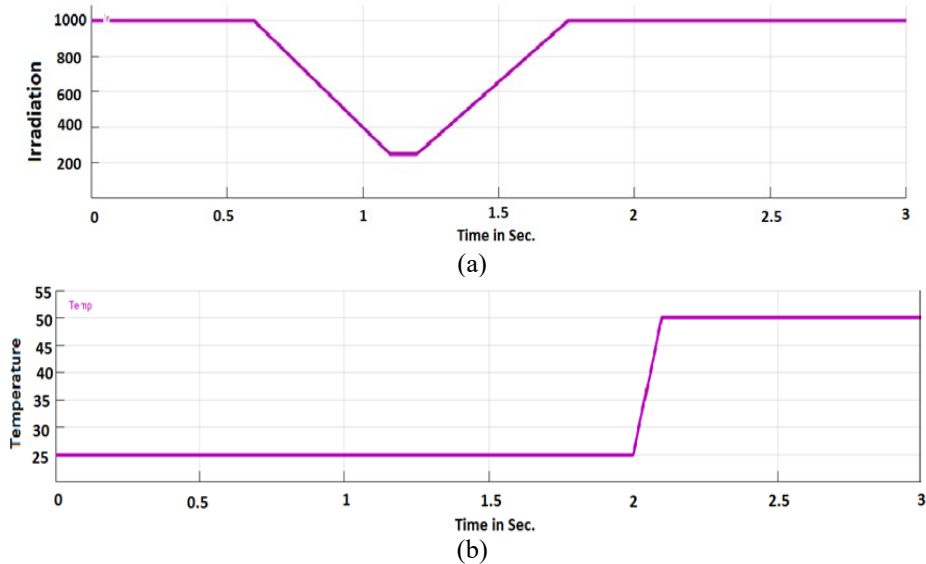


Figure 8. Weather condition [(a) Changing irradiation for PV
(b) Changing Temperature for PV array]

To prove the effectiveness of the developed controllers, the next simulations are done under the changes in Figure. 8 and 9, which in turn affect to the output of power, current and power. Figure. 8 and 9 demonstrate the operating change of the irradiation and temperature in PV module and wind speed, respectively. All simulation and controllers' parameters are given table 1 in the appendix. The next simulations are carried out via MATLAB/Simulink environment at two sides, namely, at the resources side and at the grid side.

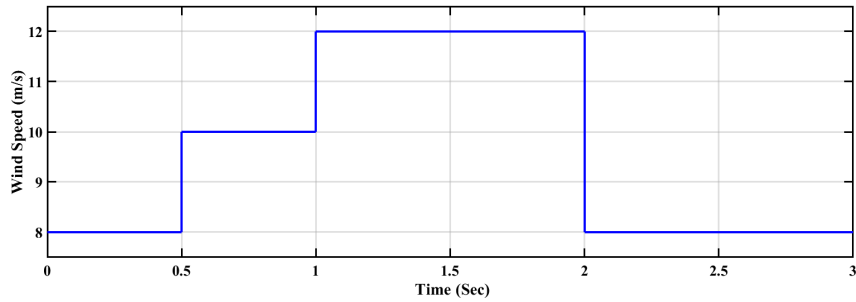


Figure 9. Wind profile for wind turbine.

A. Results at Resources Side

Considering the system demonstrated in Figure. 1, the PV currents are shown in Figure. 10 for FOPI, FIC, and classical PI controllers, respectively. The developed FOPI gives satisfactory results regarding the delivered PV current. The current from battery storage is illustrated in Figure. 11. The IC controller shows the worst performance. The battery current is steadier with the developed FOPI, which exhibits the satisfactory results. At time equals 2s, the PV temperatures increases from 25°C to 50°C, the classical PI lasts about 0.5s to restore the battery current and system stability at 2.5s. It is clear that the FOPI gives the shortest rise time with satisfactory overshoot compared to PI and FIC controllers and thus its effectiveness was affirmed.

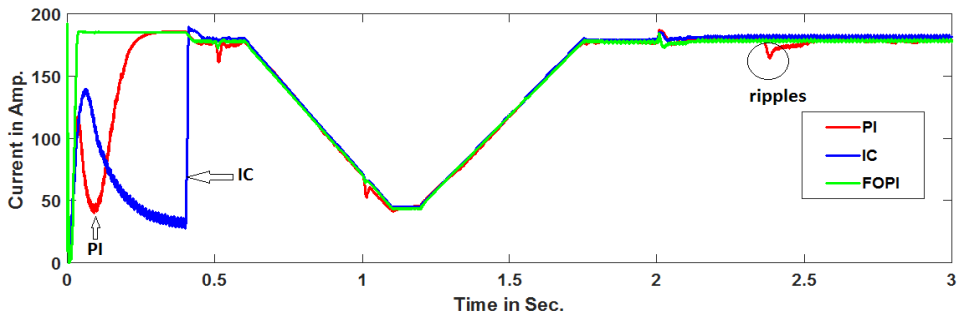


Figure 10. PV current in three types of control.

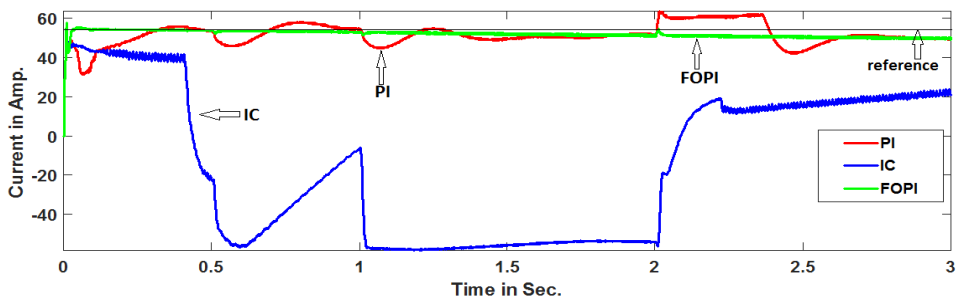


Figure 11. Battery storage current in three types of controller.

The wind turbine PMSG d- and q-axis stator currents are shown in Figure. 12 a and b for the developed controllers, respectively. At 1s, the wind speed experiences a step increase but solar irradiance continues to decrease. The IC controller demonstrates the worst performance with unsatisfactory frequency oscillations. The classic PI shows longer settling time with unsatisfactory overshoots and longer settling time (about 0.5s). The developed FOPI controller

shows satisfactory transitions with the smallest ripples at the different weather disturbances results compared to FIC and PI controllers.

Figure 13 shows the PV output of power for the developed controllers. The wind turbine output power is shown in Figure. 14. Yet, the IC shows the worst performance. The developed FOPI controller gives acceptable results with diminished ripples to track reference line compared to the IC and PI controllers and its robustness was verified.

The battery powers are shown in Figure. 15. The IC shows unsatisfactory results at the long-term operating conditions. Initially, the PI shows longer settling time with limited over/undershoots compared to the FOPI. Still, the developed FOPI controller gives satisfactory results with the shortest settling time.

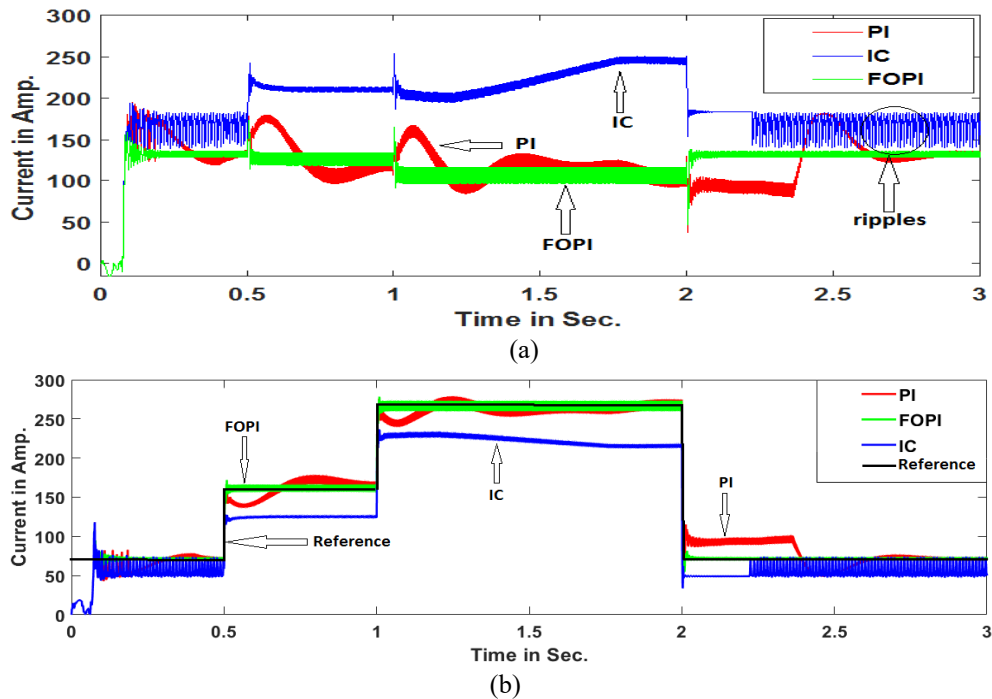


Figure 12. Stator current of wind turbine, (a) q-axis currents and (b) d-axis currents

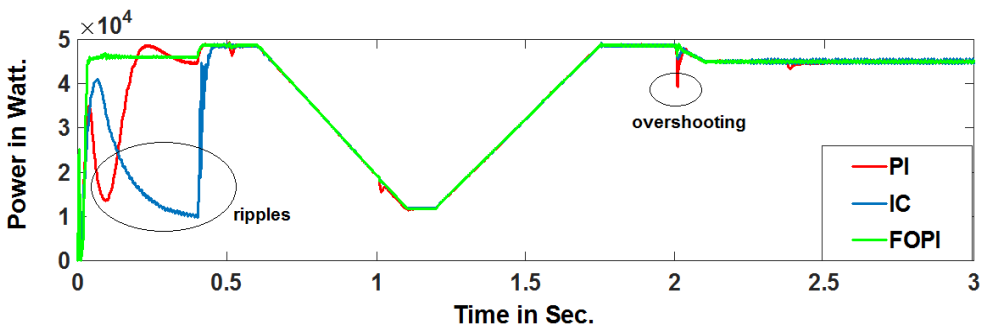
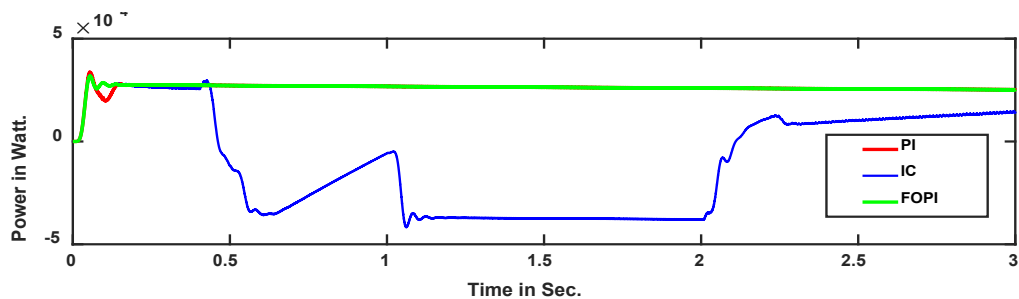
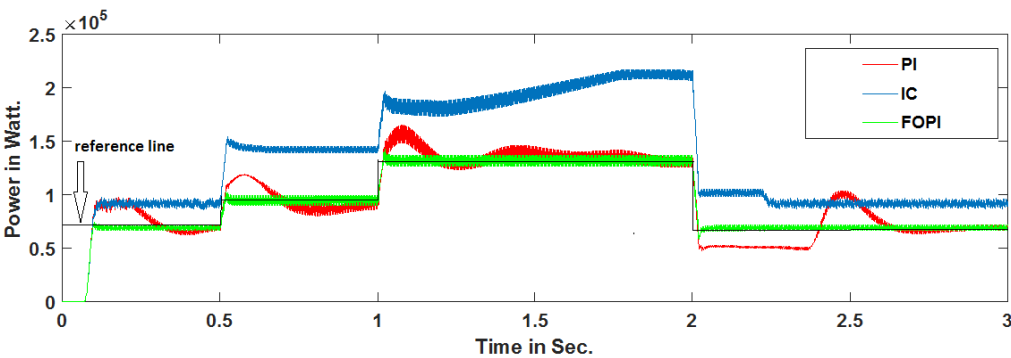
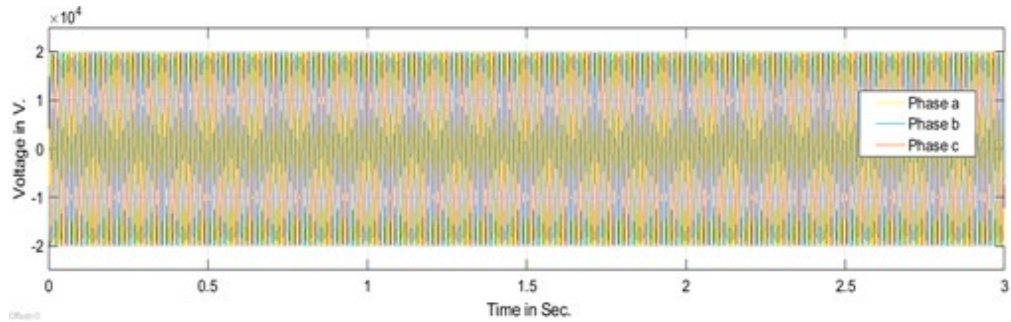


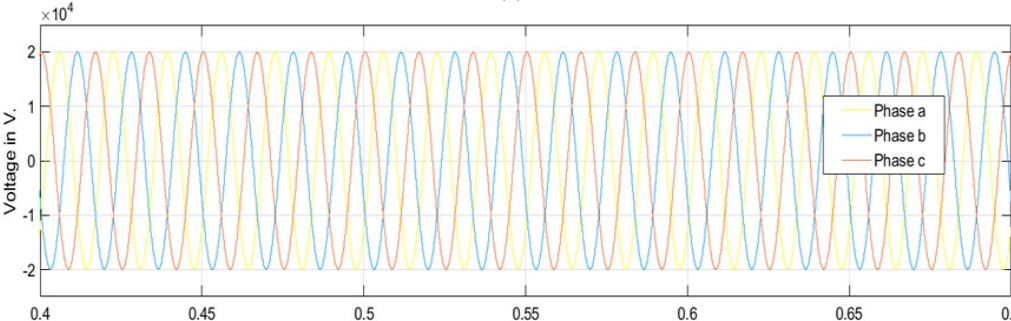
Figure 13. PV Output power.



B. Results at Grid Side



(a)



(b)

Figure 16. (a) Utility grid voltages profile and (b) Zoomed in Utility grid voltages profile

The inverter output voltage to the three-phase grid is illustrated in the Figure. 16.a and b. The grid injected current via the developed controllers is shown in Figures. 17, 18 and 19, respectively. Compared to the classical PI and IC, the developed FOPI exhibits satisfactory performance in terms of overshoots. Based on the result in Figure.19, FOPI gives the satisfactory performance, when it compared to other controllers. The FOPI controller responds quickly for the operation of inverter without ripples. Yet, FIC supplies lower current to the utility grid than other controllers, which results in smaller power delivered to the grid. IC demonstrates bad performance compared two other controllers.

The d-axis grid injected current is shown in Figures. 20. The classic PI controller tracks the reference current but at the cost of the settling time. Still the IC shows unsatisfactory results at the steady state and transient conditions. The developed FOPI shows adequate results to track the reference current and its robustness was proved.

The grid injected power via the developed PI, FOPI and IC controllers are allowed 85 kw, 86 kw and 65 kw as shown in Figure. 21. The FOPI responds satisfactorily for all changes of the wind speed and solar irradiation. However, the IC shows unsatisfactory results since it doesn't respond well for the wind speed and the irradiation changes.

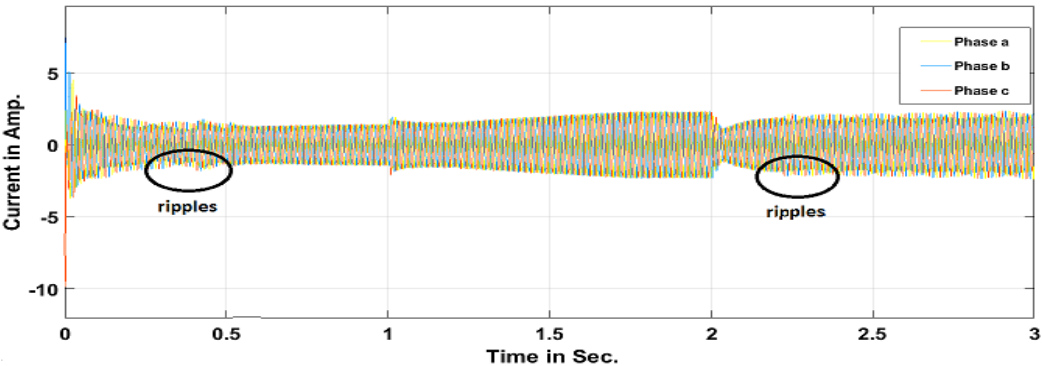


Figure 17. Grid injected current the using FIC.

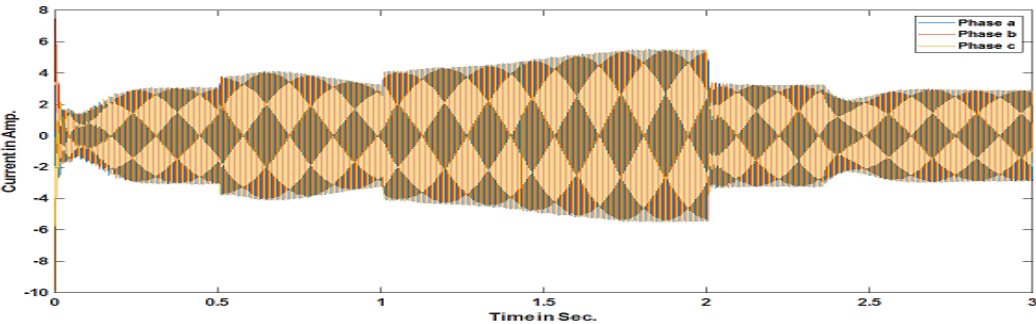


Figure 18. Grid injected current using PI controller.

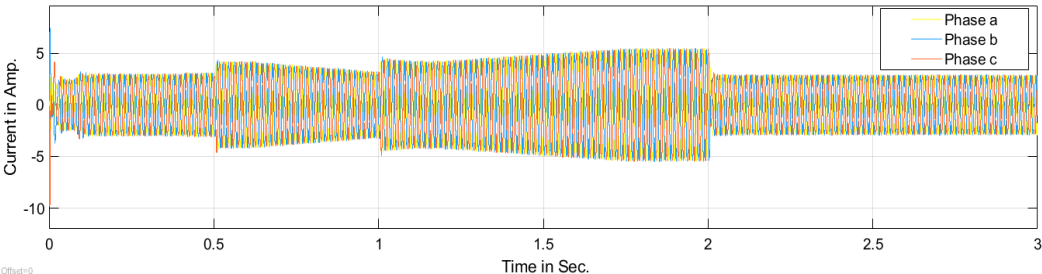
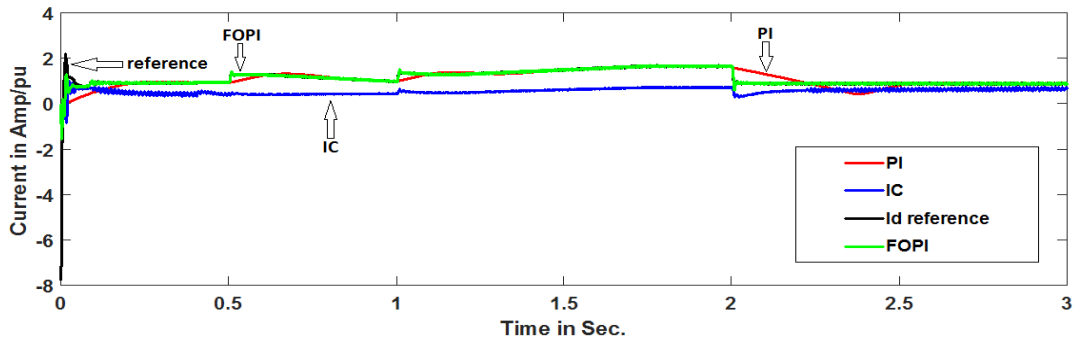
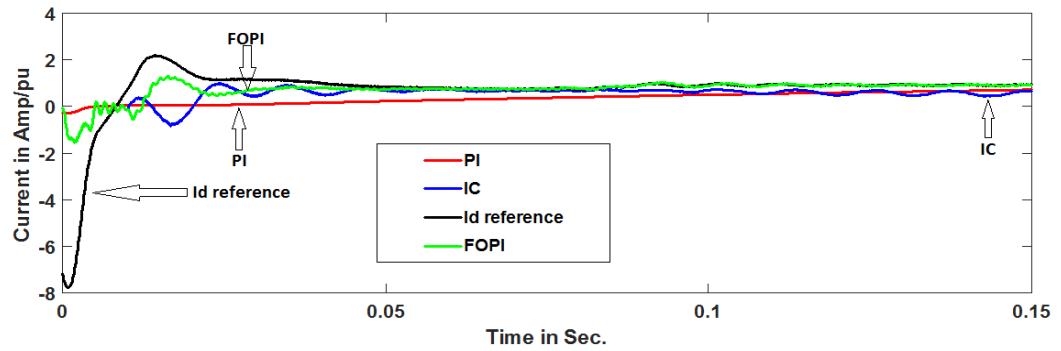


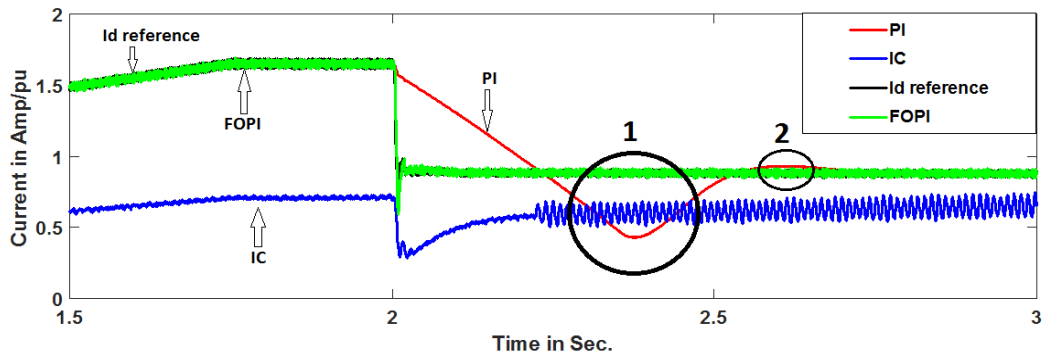
Figure 19. Grid injected current using FOPI controller.



(a)



(b)



(c)

Figure 20. (a) Injected current control, (b) Zoomed in injected current control during the first 0.15s and (c) Zoomed in injected current control at last 1.5s.

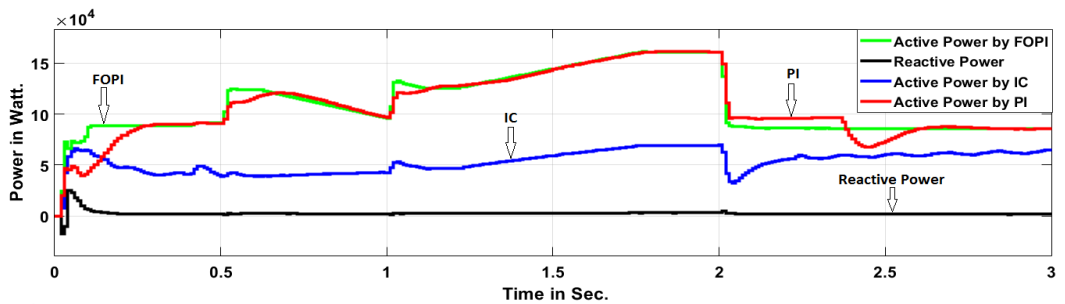


Figure 21. Grid injected power using the developed controller techniques.

5. Experimental Setup of PV-Wind Hybrid System

The data and description of the experimental system are presented table 2 in the appendix. As shown in Figure. 22, the experimental setup of a PV- wind hybrid system connected to the grid. The system consists of four busbars, the first busbar connected to the grid, the second busbar connected with loads, the third busbar connected with wind and the fourth busbar connected with PV.



Figure 22. PV- wind hybrid system with experimental work.

Figure 23 shown, the step wind speed profile varies up and down as step functions with an average value of 10 m/s with duration of 60 s.

Figure 24 shows and illustrates the change of solar radiation under the ramp profile. We observe in the form of solar radiation starting from 325 W/m² and gradually rise to 675 W/m² and then dome slightly and then the radiation gradually decreases up to 260 W/m², then stability slightly, then gradually increase to 470 W/m² and then stability slightly.

Figure 25 shows the experimental performance of PV voltage, PV current and PV power under ramp changes of solar radiation.

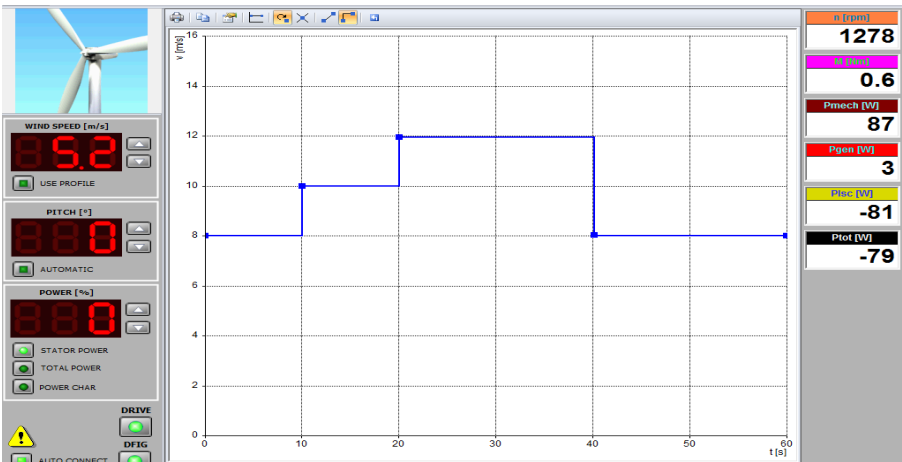


Figure 23. Wind speed profile in m/s of experimental wind energy.

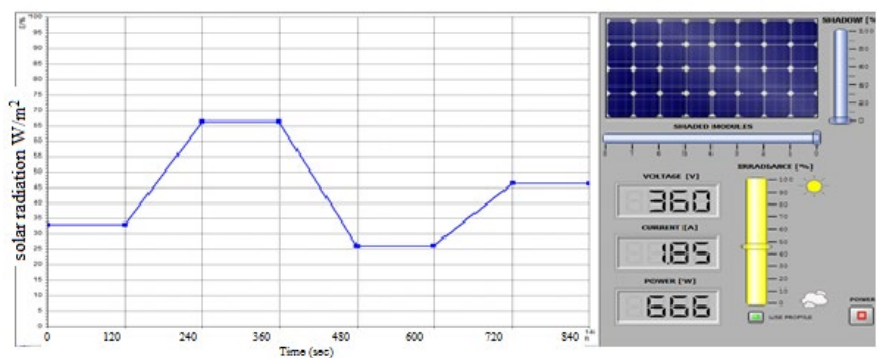


Figure 24. Ramp changes of solar radiation of experimental PV system.

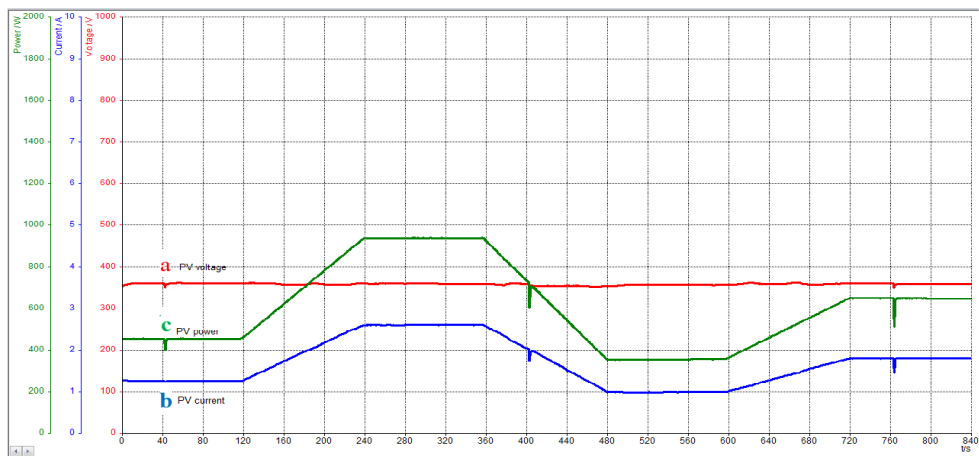


Figure 25. Experimental performance of PV voltage, PV current and PV power.

As shown in Figure. 26, the experimental performance of stator voltage in volt and stator current in ampere of wind energy. Figure 26 shows a constant voltage approximately, with a change in current, followed by a change in the wind speed.

The experimental performance of rotational wind speed of the wind generator in rpm and mechanical power in watt for wind energy, as seen in Figure 27.

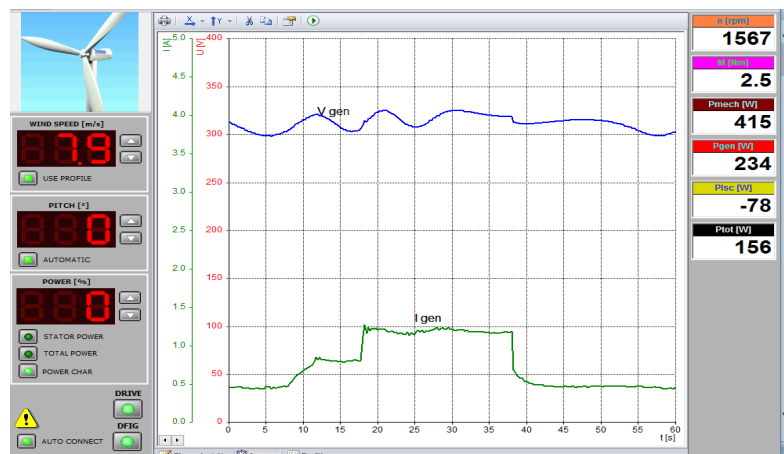


Figure 26. Experimental performance of stator voltage in volt and stator current in ampere of wind energy.

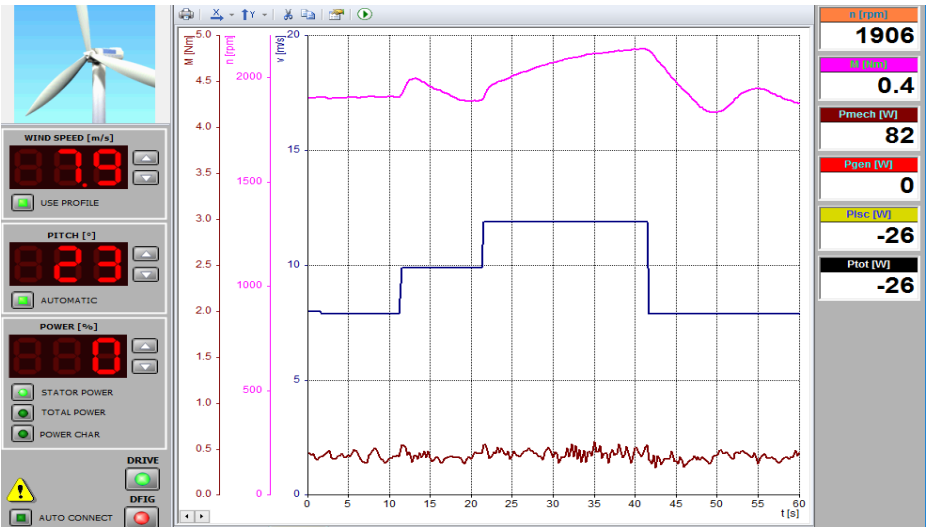


Figure 27. Experimental performance of rotational speed of the generator in rpm and mechanical power in watt.

Figure 28, describes the experimental performance of active, reactive and apparent wind power injected to the grid in watt for wind energy. A stability in the reactive power on the zero, with a change in the active power, followed by a change in the wind speed as shown in Figure. 28.

At any variation in wind speed, whether increased or decreased, followed by a similar variation in the mechanical torque and stability in the pitch angle on the zero, for the experimental response as exposed in Figure 29.

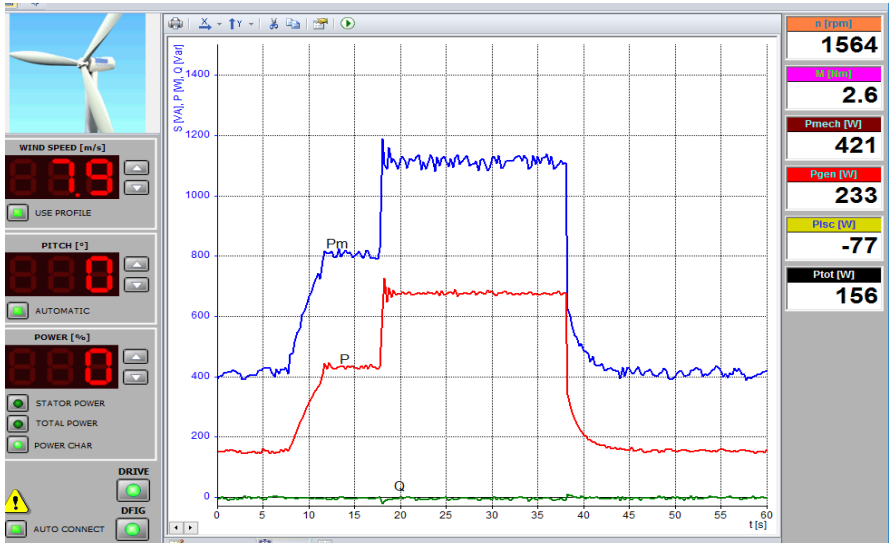


Figure 28. Experimental performance of active, reactive and apparent wind power injected to the grid in watt

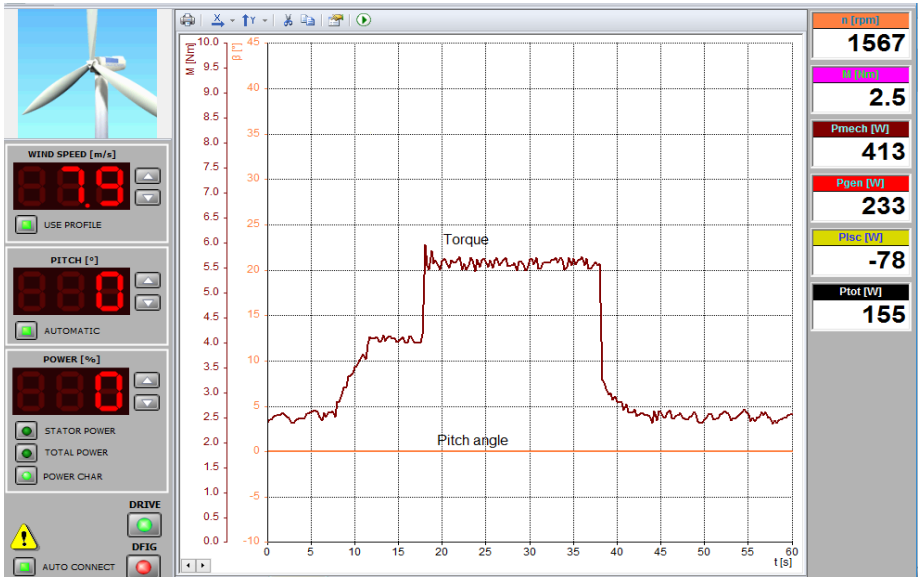


Figure 29. Experimental performance of input mechanical torque in Nm and pitch angle.

In Figure 30, it displays the results of the output power between the solar system and the wind energy connected to the grid. The output power of the wind energy was equivalent to 267 watts, while the output power of the solar energy equals 656 watts, so that the total grid connected power is 903 watts, as shown in Figure 30. In order to implement synchronization to the grid, three conditions must be same the phase sequence, grid voltage magnitude and frequency.

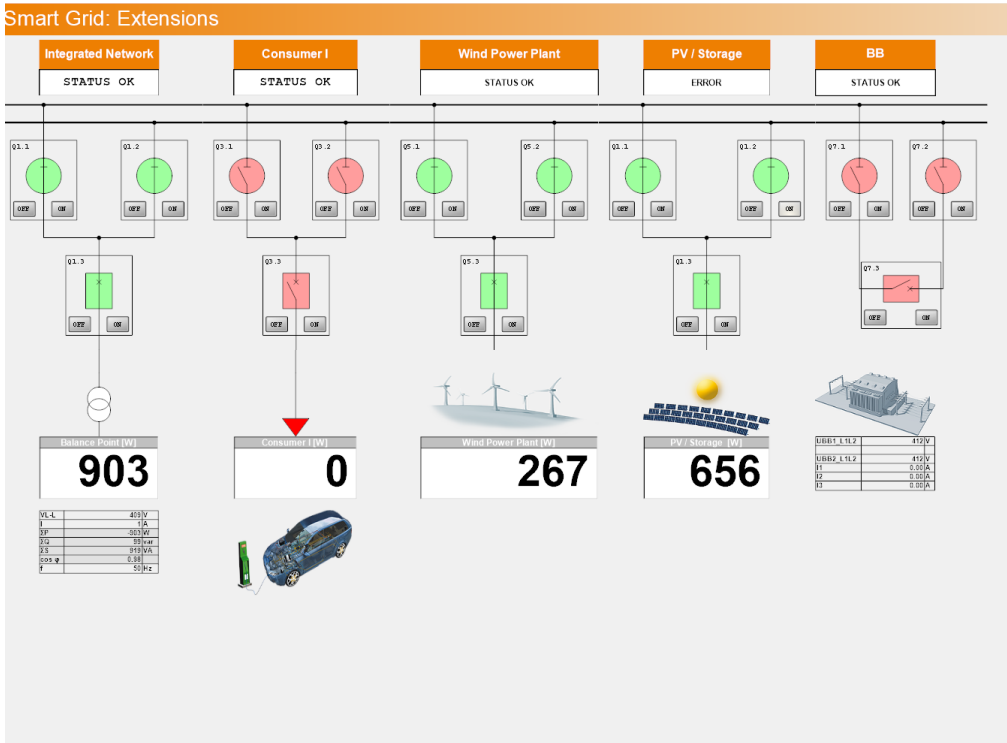


Figure 30. Experimental power results of PV – wind hybrid system connected to the grid.

6. Conclusion

This paper introduced a hybrid PV, wind turbine and battery storage system connected to a three-phase grid. In this research is compared three types of controller PI, FIC and FOPI. The effectiveness of the FOPI was demonstrated by a fair comparison with classic PI and IC. Based on digital simulation results, the following conclusions are drawn, (1) FOPI controller offers satisfactory performance to handle the power supplied to the grid, (2) FOPI can pave the way forward for more hybrid systems performance and different operating conditions, (3) IC by itself could not stabilize the system, (4) the HRES presented in this study is considered as an initial step to enhance the power provided to the grid, (5) it seems that the solar irradiation and wind speed variations have more impact upon the power delivered to the grid than the PV ambient temperature. Thus, the current work recommends utilizing the developed FOPI for future works. For the purpose of future works, techno-economic feasibility analysis could be estimated to discover the optimal size and configuration of the individual components. Furthermore, experimental analysis is to be done to highlight the stability analysis.

Through experimental results, the synchronization process between the wind energy system with PV energy and the electrical grid was verified. The results of the performance of the PV / wind synchronization with the grid are presented. It was also presented the results of the current and voltage and power PV / wind with the grid in addition to the torque-speed for standalone wind energy.

Appendix

Table 1. Simulation System Parameters.

Parameter	Value
PV Parameters	
PV maximum power of module	305.23 W
PV open circuit voltage of module	64.2 V
PV short circuit current of module	5.96 A
PV maximum voltage of module	54.7 V
PV maximum current of module	5.58 A
Parallel strings of PV array	66
Series-connected modules per string	5
Inductance of boost converter	5e-3 H
Resistance of boost converter	0.005 Ω
Capacitance of boost converter	100e-6 F
Total P_{PV} Array	45KW
Wind Parameters	
Mechanical of Wind output power	200000 W
Base wind speed	12 m/s
Maximum power at base wind speed	0.8 pu
Base power of the electrical generator	200000/0.9 VA
Pitch angle beta	0 deg
Resistance of Rectifier	100 Ohms
Capacitance of Rectifier	0.1e-6 F
P_{wind}	69.7KW
Battery Parameters	
Nominal voltage of Battery	300 V

Rated capacity	6.5 Ah
P_{Battery}	22KW
Grid Parameters	
Step up transformer	260V / 25 kV
Grid voltage	20000 V
Grid frequency	60 Hz
Controller Parameters	
Kp Current Control at PI	0.3
Ki Current Control at PI	20
Kp Voltage Source Control at PI	7
Ki Voltage Source Control at PI	800
Reference voltage of DC link	500 V
P_{out} at PI	85.35 KW
P_{out} at FOPI	86 KW
P_{out} at IC	65 KW
λ at FOPI	0.8

Table 2: Experimental setup specifications

Names	Values
PV array	
PV maximum power	1500 W
Rated current	7.2 – 11.7 A
Rated voltage	100 - 240 V
Output current	0-10 A
Output voltage	0-500 V
ON-grid inverter	
DC input voltage	250 – 1000V
DC MPP Voltage	300 – 800V
DC Max. current	11 A
DC Short circuit current	13 – 20 A
AC voltage	3*230 V
Frequency	50-60 HZ
Power Factor	0.8 – 1.0
AC Max. Current	7 A
Max. Power	3200 W
Step up transformer	
Primary Voltage	400 V
Secondary Voltage	450 V
Nominal Power	1000 VA
Frequency	50/60 HZ
Length of transmission line	150 km
Wind Energy	
Wind Unit	0.8 kW, 50 Hz, 230/400 V
Servo machine	Nominal speeds 2000 rev/min, T=6.7 N.m
Incremental encoder	6000 rpm, 1024 pulses, Moment of inertia: 35 gcm ²

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